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In re: Bantval Jayant Baliga

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For: POWER SEMICONDUCTOR DEVICES HAVING LATERALLY EXTENDING
BASE SHIELDING REGIONS THAT INHIBIT BASE REACH-THROUGH AND
SCHOTTKY RECTIFYING FLYBACK DIODES

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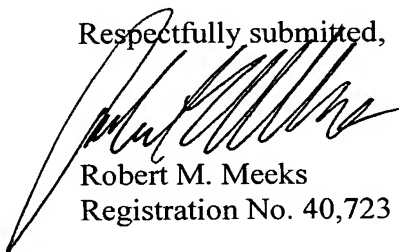
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Sir:

The above-identified application is being filed on behalf of the inventor,
Bantval Jayant Baliga, resident of Raleigh, North Carolina, under the provisions of 37 CFR
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Respectfully submitted,



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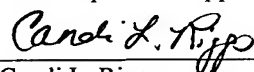
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**POWER SEMICONDUCTOR DEVICES HAVING Laterally
EXTENDING BASE SHIELDING REGIONS THAT INHIBIT BASE
REACH-THROUGH AND SCHOTTKY RECTIFYING FLYBACK DIODES**

Cross-Reference to Related Application

This application is a continuation-in-part of U.S. Application Serial No. 10/008,171, filed October 19, 2001, which is a continuation-in-part of Application Serial No. 09/833,132, filed April 11, 2001, the disclosures of which are hereby incorporated herein by reference.

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Field of the Invention

The present invention relates to semiconductor switching devices, and more particularly to switching devices for power switching and power amplification applications and methods of forming same.

Background of the Invention

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Power MOSFETs have typically been developed for applications requiring power switching and power amplification. For power switching applications, the commercially available devices are typically DMOSFETs and UMOSFETs. In these devices, one main objective is obtaining a low specific on-resistance to reduce power losses. In a power MOSFET, the gate electrode provides turn-on and turn-off control upon the application of an appropriate gate bias. For example, turn-on in an N-type enhancement MOSFET occurs when a conductive N-type inversion-layer channel (also referred to as "channel region") is formed in the P-type base region in response to the application of a positive gate bias. The inversion-layer

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channel electrically connects the N-type source and drain regions and allows for majority carrier conduction therebetween.

The power MOSFET's gate electrode is separated from the base region by an intervening insulating layer, typically silicon dioxide. Because the gate is insulated from the base region, little if any gate current is required to maintain the MOSFET in a conductive state or to switch the MOSFET from an on-state to an off-state or vice-versa. The gate current is kept small during switching because the gate forms a capacitor with the MOSFET's base region. Thus, only charging and discharging current ("displacement current") is required during switching. Because of the high input impedance associated with the insulated-gate electrode, minimal current demands are placed on the gate and the gate drive circuitry can be easily implemented. Moreover, because current conduction in the MOSFET occurs through majority carrier transport through an inversion-layer channel, the delay associated with the recombination and storage of excess minority carriers is not present. Accordingly, the switching speed of power MOSFETs can be made orders of magnitude faster than that of bipolar transistors. Unlike bipolar transistors, power MOSFETs can be designed to withstand high current densities and the application of high voltages for relatively long durations, without encountering the destructive failure mechanism known as "second breakdown". Power MOSFETs can also be easily paralleled, because the forward voltage drop across power MOSFETs increases with increasing temperature, thereby promoting an even current distribution in parallel connected devices.

DMOSFETs and UMOSFETs are more fully described in a textbook by B.J. Baliga entitled *Power Semiconductor Devices*, PWS Publishing Co. (ISBN 0-534-94098-6) (1995), the disclosure of which is hereby incorporated herein by reference. Chapter 7 of this textbook describes power MOSFETs at pages 335-425. Examples of silicon power MOSFETs including accumulation, inversion and extended trench FETs having trench gate electrodes extending into the N⁺ drain region are also disclosed in an

article by T. Syau, P. Venkatraman and B.J. Baliga, entitled *Comparison of Ultralow Specific On-Resistance UMOSFET Structures: The ACCUFET, EXTFET, INVFET, and Conventional UMOSFETs*, IEEE Transactions on Electron Devices, Vol. 41, No. 5, May (1994). As described by Syau et al., specific on-resistances in the range of 100-250 $\mu\Omega\text{cm}^2$ were experimentally demonstrated for devices capable of supporting a maximum of 25 volts. However, the performance of these devices was limited by the fact that the forward blocking voltage must be supported across the gate oxide at the bottom of the trench. U.S. Patent No. 4,680,853 to Lidow et al. also discloses a conventional power MOSFET that utilizes a highly doped N+ region **130** between adjacent P-base regions in order to reduce on-state resistance. For example, FIG. 22 of Lidow et al. discloses a high conductivity region **130** having a constant lateral density and a gradient from relatively high concentration to relatively low concentration beginning from the chip surface beneath the gate oxide and extending down into the body of the chip.

FIG. 1(d) from the aforementioned Syau et al. article discloses a conventional UMOSFET structure. In the blocking mode of operation, this UMOSFET supports most of the forward blocking voltage across the N-type drift layer, which must be doped at relatively low levels to obtain a high maximum blocking voltage capability, however low doping levels typically increase the on-state series resistance. Based on these competing design requirements of high blocking voltage and low on-state resistance, a fundamental figure of merit for power devices has been derived which relates specific on-resistance ($R_{\text{on,sp}}$) to the maximum blocking voltage (BV). As explained at page 373 of the aforementioned textbook to B.J. Baliga, the ideal specific on-resistance for an N-type silicon drift region is given by the following relation:

$$R_{\text{on,sp}} = 5.93 \times 10^{-9} (\text{BV})^{2.5} \quad (1)$$

Thus, for a device with 60 volt blocking capability, the ideal specific on-resistance is $170\ \mu\Omega\text{cm}^2$. However, because of the additional resistance contribution from the channel, reported specific on-resistances for UMOSFETs are typically much higher. For example, a UMOSFET having a specific on-resistance of $730\ \mu\Omega\text{cm}^2$ is disclosed in an article by H. Chang, entitled *Numerical and Experimental Comparison of 60V Vertical Double-Diffused MOSFETs and MOSFETs With A Trench-Gate Structure*, Solid-State Electronics, Vol. 32, No. 3, pp. 247-251, (1989). However, in this device a lower-than-ideal uniform doping concentration in the drift region was required to compensate for the high concentration of field lines near the bottom corner of the trench when blocking high forward voltages. U.S. Patent Nos. 5,637,989 and 5,742,076 and U.S. Application Serial No. 08/906,916, filed August 6, 1997, the disclosures of which are hereby incorporated herein by reference, also disclose popular power semiconductor devices having vertical current carrying capability.

In particular, U.S. Patent No. 5,637,898 to Baliga discloses a preferred silicon field effect transistor which is commonly referred to as a graded-doped (GD) UMOSFET. As illustrated by FIG. 3 from the '898 patent, a unit cell **100** of an integrated power semiconductor device field effect transistor may have a width " W_c " of $1\ \mu\text{m}$ and comprise a highly doped drain layer **114** of first conductivity type (e.g., N+) substrate, a drift layer **112** of first conductivity type having a linearly graded doping concentration therein, a relatively thin base layer **116** of second conductivity type (e.g., P-type) and a highly doped source layer **118** of first conductivity type (e.g., N+). The drift layer **112** may be formed by epitaxially growing an N-type in-situ doped monocrystalline silicon layer having a thickness of $4\ \mu\text{m}$ on an N-type drain layer **114** having a thickness of $100\ \mu\text{m}$ and a doping concentration of greater than $1 \times 10^{18}\ \text{cm}^{-3}$ (e.g. $1 \times 10^{19}\ \text{cm}^{-3}$) therein. The drift layer **112** also has a linearly graded doping concentration therein with a maximum concentration of $3 \times 10^{17}\ \text{cm}^{-3}$ at the N+/N junction with the drain layer **114**, and a minimum concentration of

1x10¹⁶ cm⁻³ beginning at a distance 3 μm from the N+/N junction (i.e., at a depth of 1 μm) and continuing at a uniform level to the upper face. The base layer **116** may be formed by implanting a P-type dopant such as boron into the drift layer **112** at an energy of 100 keV and at a dose level of 1x10¹⁴ cm⁻². The P-type dopant may then be diffused to a depth of 0.5 μm into the drift layer **112**. An N-type dopant such as arsenic may also be implanted at an energy of 50 keV and at dose level of 1x10¹⁵ cm⁻². The N-type and P-type dopants can then be diffused simultaneously to a depth of 0.5 μm and 1.0 μm, respectively, to form a composite semiconductor substrate containing the drain, drift, base and source layers.

A stripe-shaped trench having a pair of opposing sidewalls **120a** which extend in a third dimension (not shown) and a bottom **120b** is then formed in the substrate. For a unit cell **100** having a width W_c of 1 μm, the trench is preferably formed to have a width " W_t " of 0.5 μm at the end of processing. An insulated gate electrode, comprising a gate insulating region **124** and an electrically conductive gate **126** (e.g., polysilicon), is then formed in the trench. The portion of the gate insulating region **124** extending adjacent the trench bottom **120b** and the drift layer **112** may have a thickness " T_1 " of about 2000 Å to inhibit the occurrence of high electric fields at the bottom of the trench and to provide a substantially uniform potential gradient along the trench sidewalls **120a**. The portion of the gate insulating region **124** extending opposite the base layer **116** and the source layer **118** may have a thickness " T_2 " of about 500 Å to maintain the threshold voltage of the device at about 2-3 volts. Simulations of the unit cell **100** at a gate bias of 15 Volts confirm that a vertical silicon field effect transistor having a maximum blocking voltage capability of 60 Volts and a specific on-resistance ($R_{sp,on}$) of 40 μΩcm², which is four (4) times smaller than the ideal specific on-resistance of 170 μΩcm² for a 60 volt power UMOSFET, can be achieved. Notwithstanding these excellent characteristics, the transistor of FIG. 3 of the '898 patent may suffer from a relatively low high-frequency figure-of-merit (HFOM) if the overall gate-to-

drain capacitance (C_{GD}) is too large. Improper edge termination of the MOSFET may also prevent the maximum blocking voltage from being achieved. Additional UMOSFETs having graded drift regions and trench-based source electrodes are also disclosed in U.S. Patent No. 5,998,833 to
5 Baliga, the disclosure of which is hereby incorporated herein by reference.

Power MOSFETs may also be used in power amplification applications (e.g., audio or rf). In these applications the linearity of the transfer characteristic (e.g., I_d v. V_g) becomes very important in order to minimize signal distortion. Commercially available devices that are used in
10 these power amplification applications are typically the LDMOS and gallium arsenide MESFETs. However, as described below, power MOSFETs including LDMOS transistors, may have non-linear characteristics that can lead to signal distortion. The physics of current saturation in power MOSFETs is described in a textbook by S.M. Sze entitled "Physics of
15 Semiconductor Devices, Section 8.2.2, pages 438-451 (1981). As described in this textbook, the MOSFET typically works in one of two modes. At low drain voltages (when compared with the gate voltage), the MOSFET operates in a linear mode where the relationship between I_d and V_g is substantially linear. Here, the transconductance (g_m) is also
20 independent of V_g :

$$g_m = (Z/L)u_{ns}C_{ox}V_d \quad (2)$$

where Z and L are the channel width and length, respectively, u_{ns} is the
25 channel mobility, C_{ox} is the specific capacitance of the gate oxide, and V_d is the drain voltage. However, once the drain voltage increases and becomes comparable to the gate voltage (V_g), the MOSFET operates in the saturation mode as a result of channel pinch-off. When this occurs, the expression for transconductance can be expressed as:

$$g_m = (Z/L)u_{ns}C_{ox}(V_g - V_{th}) \quad (3)$$

where V_g represents the gate voltage and V_{th} represents the threshold voltage of the MOSFET. Thus, as illustrated by equation (3), during saturation operation, the transconductance increases with increasing gate bias. This makes the relationship between the drain current (on the output side) and the gate voltage (on the input side) non-linear because the drain current increases as the square of the gate voltage. This non-linearity can lead to signal distortion in power amplifiers. In addition, once the voltage drop along the channel becomes large enough to produce a longitudinal electric field of more than about 1×10^4 V/cm while remaining below the gate voltage, the electrons in the channel move with reduced differential mobility because of carrier velocity saturation.

Thus, notwithstanding attempts to develop power MOSFETs for power switching and power amplification applications, there continues to be a need to develop power MOSFETs that can support high voltages and have improved electrical characteristics, including highly linear transfer characteristics when supporting high voltages.

Summary of the Invention

Vertical power devices according to embodiments of the present invention utilize retrograded-doped transition regions to enhance forward on-state and reverse breakdown voltage characteristics. Highly doped shielding regions may also be provided that extend adjacent the transition regions and contribute to depletion of the transition regions during both forward on-state conduction and reverse blocking modes of operation.

A vertical power device (e.g., MOSFET) according to a first embodiment of the invention comprises a semiconductor substrate having first and second trenches and a drift region of first conductivity type (e.g., N-type) therein that extends into a mesa defined by and between the first and second trenches. The drift region is preferably nonuniformly doped and may have a retrograded doping profile relative to an upper surface of the substrate in which the first and second trenches are formed. In particular, the substrate may comprise a highly doped drain region of first

conductivity type and a drift region that extends between the drain region and the upper surface. The doping profile in the drift region may decrease monotonically from a nonrectifying junction with the drain region to the upper surface of the substrate and an upper portion of the drift region may be uniformly doped at a relatively low level (e.g., $1 \times 10^{16} \text{ cm}^{-3}$). First and second insulated electrodes may also be provided in the first and second trenches. These first and second insulated electrodes may constitute trench-based source electrodes in a three-terminal device.

First and second base regions of second conductivity type (e.g., P-type) are also provided in the mesa. These base regions preferably extend adjacent sidewalls of the first and second trenches, respectively. First and second highly doped source regions of first conductivity type are also provided in the first and second base regions, respectively. An insulated gate electrode is provided that extends on the mesa. The insulated gate electrode is patterned so that the upper surface preferably defines an interface between the insulated gate electrode and the first and second base regions. Inversion-layer channels are formed within the first and second base regions during forward on-state conduction, by applying a gate bias of sufficient magnitude to the insulated gate electrode.

A transition region of first conductivity type is also provided in the mesa. This transition region preferably extends between the first and second base regions and extends to the interface with the insulated gate electrode. The transition region forms a non-rectifying junction with the drift region and has a vertically retrograded first conductivity type doping profile relative to the upper surface. This doping profile has a peak doping concentration at a first depth relative to the upper surface, which may extend in a range from about 0.2 to 0.5 microns relative to the upper surface. Between the first depth and the upper surface, the doping profile is preferably monotonically decreasing in a direction towards the upper surface. A magnitude of a portion of a slope of this monotonically decreasing profile is preferably greater than $3 \times 10^{21} \text{ cm}^{-4}$. The establishment

of a "buried" peak at the first depth may be achieved by performing a single implant step at respective dose and energy levels or by performing multiple implant steps at respective dose levels and different energy levels. The peak dopant concentration in the transition region is preferably greater than at least about two (2) times the transition region dopant concentration at the upper surface. More preferably, the peak dopant concentration in the transition region is greater than about ten (10) times the transition region dopant concentration at the upper surface.

According to preferred aspects of power devices of the first embodiment, a product of the peak first conductivity type dopant concentration in the transition region (at the first depth) and a width of the transition region at the first depth is in a range between $1 \times 10^{12} \text{ cm}^{-2}$ and $7 \times 10^{12} \text{ cm}^{-2}$ and, more preferably, in a range between about $3.5 \times 10^{12} \text{ cm}^{-2}$ and about $6.5 \times 10^{12} \text{ cm}^{-2}$. Depending on unit cell design within an integrated multi-colloidal device, the product of the peak first conductivity type dopant concentration in the transition region and a width of the non-rectifying junction between the transition region and the drift region may also be in a range between $1 \times 10^{12} \text{ cm}^{-2}$ and $7 \times 10^{12} \text{ cm}^{-2}$. A product of the peak first conductivity type dopant concentration in the transition region, a width of the transition region at the first depth and a width of the mesa may also be set at a level less than $2 \times 10^{15} \text{ cm}^{-1}$. To achieve sufficient charge coupling in the drift region mesa, a product of the drift region mesa width and quantity of first conductivity type charge in a portion of the drift region mesa extending below the transition region is preferably in a range between $2 \times 10^9 \text{ cm}^{-1}$ and $2 \times 10^{10} \text{ cm}^{-1}$.

According to further aspects of the first embodiment, enhanced forward on-state and reverse blocking characteristics can be achieved by including highly doped shielding regions of second conductivity type that extend in the mesa and on opposite sides of the transition region. In particular, a first shielding region of second conductivity type is provided that extends between the first base region and the drift region and is more

highly doped than the first base region. Similarly, a second shielding region of second conductivity type is provided that extends between the second base region and the drift region and is more highly doped than the second base region. To provide depletion during forward on-state and reverse blocking modes of operation, the first and second shielding regions form respective P-N rectifying junctions with the transition region. High breakdown voltage capability may also be achieved by establishing a product of the peak first conductivity type dopant concentration in the transition region and a width between the first and second shielding regions in a range between $1 \times 10^{12} \text{ cm}^{-2}$ and $7 \times 10^{12} \text{ cm}^{-2}$.

Integrated vertical power devices according to a second embodiment of the invention preferably comprise active unit cells that provide forward on-state current and dummy cells that remove heat from the active cells during forward on-state conduction and support equivalent maximum reverse blocking voltages. According to the second embodiment, each integrated unit cell may comprise an active unit cell and one or more dummy unit cells. In addition to the first and second trenches, a third trench may be provided in the semiconductor substrate. The first and second trenches define an active mesa, in which an active unit cell is provided, and the second and third trenches define a dummy mesa therebetween in which a dummy unit cell is provided. A dummy base region of second conductivity type is provided in the dummy mesa preferably along with a dummy shielding region. The dummy base and shielding regions preferably extend across the dummy mesa and may be electrically connected to the first and second source regions within the active unit cell. In the event one or more dummy unit cells is provided, uniform reverse blocking voltage characteristics can be achieved by making the width of the mesa, in which the active unit cell is provided, equal to a width of the respective dummy mesa in which each of the dummy unit cells is provided. Alternatively, and in place of the third dummy base region, a field plate insulating layer may be provided on an upper

surface of the dummy mesa and a third insulated electrode may be provided in the third trench. The source electrode may extend on the field plate insulating layer and is electrically connected to the first, second and third insulated electrodes within the trenches. In the event a field plate
5 insulating layer is provided on the dummy mesa instead of using a dummy base region, the spacing between the first and second trenches need not necessarily equal the spacing between the second and third trenches in order to support maximum blocking voltages.

Additional embodiments of the present invention also include methods
10 of forming vertical power devices. These methods preferably include implanting transition region dopants of first conductivity type at a first dose level and first energy level into a surface of a semiconductor substrate having a drift region of first conductivity type therein that extends adjacent the surface. An insulated gate electrode may then be formed on the
15 surface. The insulated gate electrode is preferably patterned so that it extends opposite the implanted transition region dopants. Shielding region dopants of second conductivity type are then implanted at a second dose level and second energy level into the surface. This implant step is preferably performed in a self-aligned manner with respect to the gate
20 electrode, by using the gate electrode as an implant mask. Base region dopants of second conductivity type are also implanted at a third dose level and third energy level into the surface, using the gate electrode as an implant mask. Accordingly, the base and shielding region dopants are self-aligned to each other.

25 A thermal treatment step is then performed to drive the implanted transition, shielding and base region dopants into the substrate and define a transition region, first and second shielding regions on opposite sides of the transition region and first and second base regions on opposite sides of the transition region. The transition region extends into the drift region and
30 has a vertically retrograded first conductivity type doping profile therein relative to the surface. This retrograded profile is achieved by establishing

a buried peak dopant concentration sufficiently below the surface. The first and second shielding regions form respective P-N rectifying junctions with the transition region and the first and second base regions also form respective P-N rectifying junctions with the transition region. The dose and implant energies associated with the base and shielding region dopants are also selected so that the shielding regions are more highly doped relative to the base regions and extend deeper into the substrate.

According to a preferred aspect of this embodiment, the first dose and energy levels and a duration of the thermal treatment step are of sufficient magnitude that a product of a peak first conductivity type dopant concentration in the transition region and a width of the transition region, as measured between the first and second shielding regions, is in a range between $1 \times 10^{12} \text{ cm}^{-2}$ and $7 \times 10^{12} \text{ cm}^{-2}$. The first and second energy levels may also be set to cause a depth of a peak second conductivity type dopant concentration in the shielding region to be within 10% of a depth of a peak first conductivity type dopant concentration in the transition region, when the depths of the peaks are measured relative to the surface.

The step of implanting shielding region dopants is also preferably preceded by the step of forming trenches in the semiconductor substrate and lining the trenches with trench insulating layers. Conductive regions are also formed on the trench insulating layers. These trench related steps may be performed before the step of implanting the transition region dopants. In this case, the transition region dopants are preferably implanted into the conductive regions within the trenches and into mesas that are defined by the trenches. According to still further preferred aspects of this embodiment, steps are also performed to increase maximum on-state current density within the power device by improving the configuration of the source contact. In particular, the source contact is formed on a sidewall of the trenches by etching back the trench insulating layers to expose the source, base and shielding regions and then forming a source contact that ohmically contacts the conductive regions and also

contacts the source, base and shielding regions at the sidewall of each trench.

5 Vertical power MOSFETs according to further embodiments of the invention include a semiconductor substrate having a drift region of first conductivity type therein and an insulated gate electrode that extends on a first surface of the semiconductor substrate. A first base shielding region of second conductivity is provided that extends in the semiconductor substrate. The first base shielding region has a first lateral extent relative to a first end of the insulated gate electrode. A first base region of second conductivity type is also provided in the substrate. The first base region extends between the first base shielding region and the first surface. The first base region has a second lateral extent relative to the first end of the insulated gate electrode that is less than the first lateral extent. The power device also includes a first source region of first conductivity type that extends in and forms a P-N junction with the first base region. A transition region of first conductivity type is provided that extends between the drift region and a portion of the first surface extending opposite the insulated gate electrode. The transition region forms rectifying junctions with the first base region and the first base shielding region. An upper portion of the transition region has a vertically retrograded first conductivity type doping profile. The vertically retrograded first conductivity type doping profile may have a peak at a first depth relative to the first surface.

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A second base region and a second base shielding region may also be provided in the substrate. In particular, the first and second base regions may be self-aligned to first and second opposing ends of the insulated gate electrode and may form respective P-N junctions with opposing sides of an upper portion of the transition region extending adjacent the first surface. The first and second base shielding regions are more highly doped than the first and second base regions and extend laterally towards each other in the semiconductor substrate to thereby constrict a neck of the upper portion of said transition region to a minimum

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width at a second depth relative to the first surface. The second depth is preferably greater than about 0.25 microns. A product of the peak first conductivity type dopant concentration in the transition region and a width of the transition region at the first depth is preferably in a range between about $1 \times 10^{12} \text{ cm}^{-2}$ and about $7 \times 10^{12} \text{ cm}^{-2}$, and more preferably in a range between about $3.5 \times 10^{12} \text{ cm}^{-2}$ and about $6.5 \times 10^{12} \text{ cm}^{-2}$.

Methods of forming these vertical MOSFETs may include forming a semiconductor substrate having a drift region of first conductivity type therein and a transition region of first conductivity type that extends between the drift region and a first surface of the semiconductor substrate. A gate electrode is then formed on the first surface. After the gate electrode has been formed, base shielding region dopants of second conductivity type are implanted at a relatively high dose and high energy level into an upper portion of the transition region, using the gate electrode as an implant mask. The peak concentration of implanted base shielding region dopants is sufficiently spaced from the first surface that buried base shielding regions can be formed with the characteristics described herein. The semiconductor substrate is then annealed to partially drive the base shielding region dopants vertically into the transition region and laterally underneath the gate electrode. This annealing step results in the definition of first and second intermediate shielding regions. Base region dopants of second conductivity type are then implanted at a relatively low dose and low energy level into upper portions of the first and second intermediate shielding regions. During this implant step, the gate electrode is used again as an implant mask in order to provide a self-aligned feature. Another annealing step is then performed to drive the base region dopants vertically into the substrate and laterally along the first surface and underneath the gate electrode to thereby define first and second base regions. During this annealing step, the base shielding region dopants are also driven laterally and vertically to substantially their full and final depth within the substrate. Based on the early implant and multiple annealing

steps, first and second base shielding regions are defined that constrict a neck of the upper portion of the transition region to a minimum width at a depth corresponding to the depth at which the original peak concentration of implanted base shielding region dopants is achieved. First and second source regions are then formed in the first and second base regions, respectively.

Power devices according to still further embodiments of the present invention include a semiconductor substrate having a drift region of first conductivity type therein and transition region of first conductivity type that extends between the drift region and a first surface of the semiconductor substrate. This transition region has a vertically retrograded first conductivity type doping profile therein that peaks at a first depth relative to the first surface. First and second shielding regions of second conductivity type are also provided. These shielding regions extend in the drift region and define respective P-N junctions with the transition region. In particular, the first and second shielding regions extend laterally towards each other in a manner that constricts a neck of the transition region to a minimum width at a second depth relative to the first surface. An anode electrode may also be provided on the first surface of the semiconductor substrate. This anode electrode defines a Schottky rectifying junction with the transition region. According to preferred aspects of these embodiments, the transition region is designed so that a product of the peak first conductivity type dopant concentration in the transition region and a width of the transition region at the first depth is in a range between about $1 \times 10^{12} \text{ cm}^{-2}$ and about $7 \times 10^{12} \text{ cm}^{-2}$ and, more preferably, in a range between about $3.5 \times 10^{12} \text{ cm}^{-2}$ and about $6.5 \times 10^{12} \text{ cm}^{-2}$.

Brief Description of the Drawings

FIG. 1 is a cross-sectional view of a vertical power device according to a first embodiment of the present invention.

FIG. 2 is a cross-sectional view of a vertical power device according to a second embodiment of the present invention.

FIG. 3 is a cross-sectional view of a vertical power device according to a third embodiment of the present invention.

FIG. 4 is a cross-sectional view of a vertical power device according to a fourth embodiment of the present invention.

5 FIG. 5 is a cross-sectional view of a vertical power device according to a fifth embodiment of the present invention.

FIG. 6 is a cross-sectional view of a vertical power device according to a sixth embodiment of the present invention.

10 FIG. 7 is a cross-sectional view of a vertical power device according to a seventh embodiment of the present invention.

FIG. 8A is a graphical illustration of a preferred vertically retrograded doping profile across the transition region of the embodiment of FIG. 1, obtained by performing multiple implants of transition region dopants at respective different energies.

15 FIG. 8B is a graphical illustration of a preferred vertical doping profile across the source, base and shielding regions of the embodiment of FIG. 1.

FIGS. 9A-9K are cross-sectional views of intermediate structures that illustrate preferred methods of forming the vertical power device of FIG. 5.

20 FIG. 10 is a cross-sectional view of a vertical power device according to another embodiment of the present invention.

FIG. 11 is a cross-sectional view of a vertical power device that includes a dummy gate electrode electrically connected to a source electrode, according to another embodiment of the present invention.

25 FIG. 12 is a cross-sectional view of a conventional double-diffused power MOSFET.

FIG. 13 is a cross-sectional view of a vertical power device unit cell according to another embodiment of the present invention.

FIGS. 14A-14G are cross-sectional views of intermediate structures that illustrate methods of forming the device of FIG. 13.

FIG. 15 is a cross-sectional view of a vertical power device with Schottky rectifying flyback diode, according to another embodiment of the present invention.

Description of Preferred Embodiments

5 The present invention will now be described more fully hereinafter with reference to the accompanying drawings, in which preferred embodiments of the invention are shown. This invention may, however, be embodied in different forms and should not be construed as limited to the embodiments set forth herein. Rather, these embodiments are provided so
10 that this disclosure will be thorough and complete, and will fully convey the scope of the invention to those skilled in the art. In the drawings, the thickness of layers and regions are exaggerated for clarity. It will also be understood that when a layer is referred to as being "on" another layer or substrate, it can be directly on the other layer or substrate, or intervening
15 layers may also be present. Moreover, the terms "first conductivity type" and "second conductivity type" refer to opposite conductivity types such as N or P-type, however, each embodiment described and illustrated herein includes its complementary embodiment as well. Like numbers refer to like elements throughout.

20 Referring now to FIG. 1, an integrated vertical power device **10** according to a first embodiment of the present invention includes a plurality of active vertical power device unit cells located side-by-side in a semiconductor substrate. As illustrated, the power device **10** comprises a highly doped drain region **100** of first conductivity type (shown as N+) and a
25 drift region **102** of first conductivity type that forms a non-rectifying junction with the drain region **100**. A drain electrode **136** is also provided in ohmic contact with the drain region **100**. The drain region **100** may have a thickness in a range between about 10 microns and about 500 microns. The drift region **102** is preferably nonuniformly doped. In particular, the
30 drift region **102** preferably has a graded doping profile which decreases monotonically in a direction extending from the non-rectifying junction to a

first surface **102a** of the drift region **102**. This graded doping profile may be a linearly graded doping profile that decreases from a preferred maximum drift region dopant concentration in a range between about 1×10^{17} and about $2.5 \times 10^{17} \text{ cm}^{-3}$ to a minimum dopant concentration. Accordingly, if the drain region **100** is doped at a level of about $1 \times 10^{19} \text{ cm}^{-3}$ or greater, then the non-rectifying junction will be an abrupt non-rectifying junction. An upper portion of the drift region **102** may be uniformly doped at a level of about $1 \times 10^{16} \text{ cm}^{-3}$ and the uniformly doped upper portion of the drift region **102** may have a thickness in a range between about 0.5 and about 1.0 μm .

A plurality of trenches **104** may be formed in the drift region **102**. If trenches are provided, the trenches **104** are preferably formed side-by-side in the drift region **102** as parallel stripe-shaped trenches, however, other trench shapes (e.g., arcuate, serpentine and polygon, including ring and hex-shaped, etc.) may also be used. As described herein, regions will be defined as separate regions if they appear as such when viewed in transverse cross-section. Each pair of trenches preferably defines a drift region mesa **102b** therebetween, as illustrated. An electrically insulating layer **106** is also provided on the sidewalls and bottoms of the trenches **104**. The "trench" insulating layer **106** may have a thickness of about 3000 Å, however, the thickness may vary depending, among other things, on the rating of the power device **10**. The electrically insulating layer **106** may comprise silicon dioxide or another conventional dielectric material. Each of the trenches **104** is preferably filled with a conductive region **110** that is electrically insulated from the drift region **102** by a respective electrically insulating layer **106**. The conductive regions **110** may constitute trench-based electrodes that are electrically connected together by a source electrode **138**. This source contact/electrode **138** may extend on the first surface **102a** of the drift region **102**, as illustrated.

Upper uniformly doped portions of the drift region mesas **102b** preferably comprise respective transition regions **130** of first conductivity type. The transition regions **130** form respective non-rectifying junctions

with the drift region **102** and, depending on thickness, may form respective non-rectifying junctions with the uniformly doped upper portions of the drift region **102** or the graded doped portions of the drift region **102**. For example, the uniformly doped upper portions of the drift region **102** may have a thickness of about 1.0 μm relative to the first surface **102a** and the transition regions **130** may have thicknesses of about 0.7 μm relative to the first surface **102a**. Moreover, according to a preferred aspect of the present invention, each transition region **130** has a first conductivity type doping profile therein that is vertically retrograded relative to the first surface **102a**. In particular, a peak first conductivity type dopant concentration at a first depth in the transition region is at least two (2) times greater than a value of the retrograded first conductivity type doping profile at the first surface **102a**. More preferably, the peak first conductivity type dopant concentration in the transition region is at least about ten (10) times greater than the value of the first conductivity type dopant concentration at the first surface. According to another preferred aspect, a slope of at least a portion of the retrograded first conductivity type doping profile is greater than about $3 \times 10^{21} \text{ cm}^{-4}$. The doping profile in the transition region **130** also includes a high-to-low graded profile in a direction extending downward from the peak to the non-rectifying junction between the transition region **130** and the drift region **102**. A desired doping profile may be achieved by performing a single transition region implant step at relatively high energy and dose or performing multiple implant steps. For example, as illustrated by FIG. 8A, a relatively wide peak in the transition region doping profile may be achieved by performing three implant steps at respective energies (and same or similar dose levels) to achieve first, second and third implant depths of about 0.15, 0.3 and 0.45 microns, using a dopant having a characteristic diffusion length of about 0.1 microns.

Gate electrodes **118** are provided on the first surface **102a**, as illustrated. These gate electrodes **118** may be stripe-shaped and may extend parallel to the trench-based electrodes **110**. As illustrated, the gate

electrodes **118** preferably constitute insulated gate electrodes (e.g., MOS gate electrodes). The gate electrodes **118** may also extend in a lengthwise direction that is orthogonal to the lengthwise direction of the trench-based electrodes **110**, with the illustrated regions **130**, **133**, **126** and **128** within each drift region mesa **102b** being rotated 90° relative to the sidewalls of the trenches **104**. The vertical power device **10** also comprises highly doped shielding regions **128** of second conductivity type (shown as P+) that are formed at spaced locations in the drift region mesas **102b**. These shielding regions **128** are preferably self-aligned to the gate electrodes **118**. Each of the shielding regions **128** preferably forms a P-N rectifying junction with a respective side of the transition region **130** and with a respective drift region mesa **102b** (or tail of the transition region **130**). According to a preferred aspect of the present invention, the peak second conductivity type dopant concentration in each shielding region **128** is formed at about the same depth (relative to the first surface **102a**) as the peak first conductivity type dopant concentration in a respective transition region **130**. Base regions **126** of second conductivity type (shown as P) are also formed in respective drift region mesas **102b**. Each base region **126** is preferably self-aligned to a respective gate electrode **118**. Highly doped source regions **133** of first conductivity type (shown as N+) are also formed in respective base regions **126**, as illustrated. The spacing along the first surface **102a** between a source region **133** and a respective edge of the transition region **130** defines the channel length of the power device **10**. These source regions **133** ohmically contact the source electrode **138**. Edge termination may also be provided by extending the source electrode **138** over peripheral drift region extensions **102c** and by electrically isolating the source electrode **138** from the peripheral drift region extensions **102c** by a field plate insulating region **125**.

The combination within each drift region mesa **102b** of (i) a pair of spaced-apart shielding regions **128** and (ii) a preferred transition region **130** that extends between the shielding regions **128** and has a vertically

retrograded doping profile, can enhance the breakdown voltage characteristics of each active unit cell in the multi-celled power device **10**. In particular, the shielding regions **128** can operate to “shield” the respective base regions **126** by significantly suppressing P-base reach-through effects when the power device **10** is blocking reverse voltages and by causing reverse current to flow through the shielding regions **128** instead of the base regions **126**. This suppression of P-base reach-through enables a reduction in the channel length of the device **10**. Moreover, the preferred retrograded doping profile in the transition region **130** enables complete or full depletion of the transition region **130** when the power device **10** is blocking maximum reverse voltages and the drift region mesa **102b** is supporting the reverse voltage.

Full depletion of the transition region **130** may also occur during forward on-state conduction. In particular, full depletion during forward operation preferably occurs before the voltage in the channel (at the end adjacent the transition region **130**) equals the gate voltage on the insulated gate electrode **118**. As used herein, the reference to the transition region being “fully depleted” should be interpreted to mean that the transition region is at least sufficiently depleted to provide a JFET-style pinch-off of a forward on-state current path that extends vertically through the transition region **130**. To achieve full depletion, the relatively highly doped shielding regions **128** of second conductivity (e.g., P+) are provided in close proximity and on opposite sides of the transition region **130**. As the voltage in the channel increases during forward on-state conduction, the transition region **130** becomes more and more depleted until a JFET-style pinch-off occurs within the transition region **130**. This JFET-style pinch-off in the transition region **130** can be designed to occur before the voltage at the drain-side of the channel (V_{cd}) equals the gate voltage (i.e., $V_{cd} \leq V_{gs}$). For example, the MOSFET may be designed so that the transition region **130** becomes fully depleted when $0.1 \leq V_{cd} \leq 0.5$ Volts and $V_{gs} = 4.0$ Volts. Use of the preferred transition region **130** enables the field effect transistor within

the power device **10** to operate in a linear mode of operation during forward on-state conduction while a drain region of the transistor simultaneously operates in a velocity saturation mode of operation. Other power devices that exhibit similar modes of operation are described in U.S. Application
5 Serial No. 09/602,414, filed June 23, 2000, entitled "MOSFET Devices Having Linear Transfer Characteristics When Operating in Velocity Saturation Mode and Methods of Forming and Operating Same", assigned to the present assignee, the disclosure of which is hereby incorporated herein by reference.

10 Simulations of the device of FIG. 1 were also performed for a unit cell having a trench depth of 4.7 microns, a trench width of 1.1 microns and a mesa width of 1.9 microns. A sidewall oxide thickness of 3000Å was also used. The drift region had a thickness of 6 microns and the uniformly doped upper portion of the drift region had a thickness of 0.5 microns. The
15 concentration of first conductivity type dopants in the uniformly doped upper portion of the drift region was set at $1 \times 10^{16} \text{ cm}^{-3}$ and the drain region had a phosphorus doping concentration of $5 \times 10^{19} \text{ cm}^{-3}$. The gate oxide thickness was set at 250Å and a total gate length (across the mesa) of 0.9 microns was used. The widths of the shielding, base and source regions
20 (relative to the sidewalls) were 0.65, 0.65 and 0.45 microns, respectively, and the channel length was 0.2 microns. The width of the transition region (at the depth of the peak concentration in the transition region) was set at 0.6 microns. The depths of the source, base, shielding and transition regions and their peak dopant concentrations can be obtained from the
25 following Table 1 and FIGS. 8A-8B, where Peak N_d and Peak N_a are the peak donor and acceptor concentrations.

5

Region	Implant Energy (KeV)	Implant Dose (cm ⁻²)	Dopant	Peak N _{d,a} cm ⁻³
N+ source	40-50	1-5 x 10 ¹⁵	P, As	1 x 10 ²⁰
P-base	40-50	1-5 x 10 ¹³	B	2 x 10 ¹⁸ (surface); 4 x 10 ¹⁷ (channel max)
P+ shield	100	1-5 x 10 ¹⁴	B	5 x 10 ¹⁸
-transition	200	1-10 x 10 ¹²	P	1.3 x 10 ¹⁷

TABLE 1

Based on the above characteristics and including variations of the peak dopant concentration in the transition region (Peak_{TR}) and width of the transition region (W_{TR}), the following simulated breakdown voltages of Tables 2 and 3 were obtained. Medici™ simulation software, distributed by Avant!™ Corporation, was used to perform the device simulations.

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W _{TR} (μm)	(Peak _{TR})(cm ⁻³)	BV (Volts)	Q(#/cm ²)
0.5	0.4 x 10 ¹⁷	80	0.2 x 10 ¹³
0.5	0.7 x 10 ¹⁷	80	0.35 x 10 ¹³
0.5	1.2 x 10 ¹⁷	79	0.6 x 10 ¹³
0.5	1.3 x 10 ¹⁷	78	0.65 x 10 ¹³
0.5	1.4 x 10 ¹⁷	62	0.7 x 10 ¹³
0.5	1.6 x 10 ¹⁷	35	0.8 x 10 ¹³
0.5	1.9 x 10 ¹⁷	20	0.95 x 10 ¹³
0.5	2.5 x 10 ¹⁷	9	1.25 x 10 ¹³

25

TABLE 2

5

$W_{TR}(\mu m)$	$(Peak_{TR})(cm^{-3})$	BV (Volts)	$Q(\#/cm^2)$
0.3	1.4×10^{17}	80	0.42×10^{13}
0.4	1.4×10^{17}	80	0.56×10^{13}
0.5	1.4×10^{17}	62	0.7×10^{13}
0.6	1.4×10^{17}	37	0.84×10^{13}
0.7	1.4×10^{17}	24	0.98×10^{13}

TABLE 3

10 As determined by the inventor herein and illustrated by the simulation results of Tables 2 and 3, power devices having high breakdown voltages can be provided by establishing a product of the peak first conductivity type dopant concentration in the transition region (at the first depth) and a width of the transition region at the first depth in a preferred range that is

15 between about $1 \times 10^{12} \text{ cm}^{-2}$ and about $7 \times 10^{12} \text{ cm}^{-2}$ and, more preferably, in a range between about $3.5 \times 10^{12} \text{ cm}^{-2}$ and about $6.5 \times 10^{12} \text{ cm}^{-2}$. This narrower more preferred range can result in devices having high breakdown voltage and excellent on-state resistance characteristics.

20 Depending on unit cell design within an integrated multi-celled device, the product of the peak first conductivity type dopant concentration in the transition region and a width of the non-rectifying junction between the transition region and the drift region may also be in a range between about $1 \times 10^{12} \text{ cm}^{-2}$ and about $7 \times 10^{12} \text{ cm}^{-2}$. A product of the peak first conductivity type dopant concentration in the transition region, a width of the transition

25 region at the first depth and a width of the mesa may also be set at a level less than about $2 \times 10^{15} \text{ cm}^{-1}$. To achieve sufficient charge coupling in the drift region mesa, a product of the drift region mesa width and quantity of first conductivity type charge in a portion of the drift region mesa extending below the transition region is preferably in a range between about 2×10^9

30 cm^{-1} and about $2 \times 10^{10} \text{ cm}^{-1}$.

Referring now to FIGS. 2-7, additional embodiments of power devices according to the present invention include the multi-celled power device **20** of FIG. 2. This device **20** is similar to the device **10** of FIG. 1, however, antiparallel diodes are provided by Schottky rectifying contacts that extend between the source electrode **138** and the drift region extensions **102c**.
The power device **30** of FIG. 3 is also similar to the power device **20** of FIG. 2, however, a plurality of dummy unit cells are provided in dummy drift region mesas **102d**. Dummy shielding regions (shown as P+) and dummy base regions (shown as P) are also provided in the dummy drift region mesas **102d**. As illustrated, the dummy base regions electrically contact the source electrode **138**. The dummy base regions and dummy shielding region can be formed at the same time as the base and shielding regions within the active unit cells. Depending on the thermal ratings of a multi-celled power device, one or more dummy unit cells may be provided to facilitate heat removal from each active unit cell.

The multi-celled power device **40** of FIG. 4 is similar to the device **30** of FIG. 3, however, the dummy drift region mesas **102d** (which may not contribute to forward on-state conduction, but preferably support equivalent reverse breakdown voltages) are capacitively coupled through a field plate insulating layer **125** to the source electrode **138**. In contrast to the widths of the dummy drift region mesas **102d** in FIG. 3, which should be equal to the widths of the drift region mesas **102b** of the active unit cells, the widths of the dummy drift region mesas **102d** in FIG. 4 need not be equal. The power device **50** of FIG. 5 is similar to the device **20** of FIG. 2, however, the electrically insulating layers **106** on the sidewalls of the trenches have been recessed to enable direct sidewall contact between the source electrode **138** and the source, base and shielding regions within the active unit cells. The establishment of this direct sidewall contact increases the active area of the device **50** by reducing and preferably eliminating the requirement that the source regions be periodically interrupted in a third

dimension (not shown) in order to provide direct contacts to the base regions.

5 The power device **60** of FIG. 6 illustrates a relatively wide active drift region mesa **102b** with a centrally located base region **126a** and shielding region **128a**. The transition region **130a** may have the same characteristics as described above with respect to the transition regions **130** within the power devices **10-50** of FIGS. 1-5. The power device **70** of FIG. 7 is similar to the device **60** of FIG. 6, however, the centrally located base region **126a** and shielding region **128a** of FIG. 6 have been separated by a
10 centrally located trench **104**. The power device **10'** of FIG. 10 is similar to the power device **10** of FIG. 1, however, the insulated gate electrode **118** on each active mesa **102b** has been replaced by a pair of shorter insulated gate electrodes **118a** and **118b**. For a mesa having a width of 2.6 microns, the gate electrodes **118a** and **118b** may have a length of 0.3 microns, for
15 example. The use of a pair of shorter gate electrodes instead of a single continuous gate electrode that extends opposite the entire width of the transition region **130** can reduce the gate-to-drain capacitance C_{gd} of the device **10'** and increase high frequency power gain. The source electrode **138** also extends into the space between the gate electrodes **118a** and
20 **118b**, as illustrated by Fig. 10. The portion of the source electrode **138** that extends into the space between the gate electrodes **118a** and **118b** may have a length of about 0.2 microns. The insulator that extends directly between the source electrode **138** and the transition region **130** may be a gate oxide and may have a thickness in a range between about 100 Å and
25 about 1000Å. The sidewall insulator that extends between the sidewalls of the gate electrodes **118a** and **118b** and the source electrode **138** may also have a thickness in a range between about 1000 Å and about 5000 Å, however, other sidewall insulator thicknesses may also be used. According to another aspect of this embodiment, the portion of the source electrode
30 **138** that extends into the space between the gate electrodes **118a** and **118b** may be formed by patterning a conductive layer (e.g., polysilicon)

used to form the gate electrode **118a** and **118b**. In particular, a third "dummy" gate electrode **118c** may be patterned that extends opposite the transition region **130**. An illustration of a vertical power device **10''** that utilizes a dummy gate electrode **118c** is provided by FIG. 11. The device
5 **10''** of FIG. 11 may otherwise be similar to the device **10'** of FIG. 10. Electrical contact between this third dummy gate electrode **118c** and the source electrode **138** may be made using conventional back-end processing techniques.

Preferred methods of forming the vertical power device of FIG. 5 with
10 a 65 Volt product rating will now be described. As illustrated by FIG. 9A, these methods may include the step of epitaxially growing a drift region **202** of first conductivity type (shown as N) on a highly doped silicon substrate **200** (e.g., N+ substrate). This highly doped substrate **200** may have a first conductivity type doping concentration therein of greater than about 1×10^{19}
15 cm^{-3} and may have an initial thickness T_s of about 500 microns. The epitaxial growth step is preferably performed while simultaneously doping the drift region **202** with first conductivity type dopants in a graded manner. To achieve a 65 Volt product rating, a vertical power device having an actual blocking voltage of 75 Volts may be required. To achieve this
20 blocking voltage, trenches having a depth in a range between about 4.5-5 microns will typically be required. To support trenches with this depth, a graded doped drift region **202** having a thickness T_d of about 6 microns may be required. Preferably, a drift region **202** having a thickness of 6 microns will include a uniformly doped region at an upper surface thereof.
25 This uniformly doped region may have a thickness in a range between about 0.5 and 1.0 microns and may be doped at a uniform level of about $1 \times 10^{16} \text{ cm}^{-3}$. The graded-doped portion of the drift region **202** may have a thickness of 5.0-5.5 microns and may be graded from a doping level of $1 \times 10^{16} \text{ cm}^{-3}$ at a depth of 0.5 or 1.0 microns, for example, to a higher level
30 of at least about $5 \times 10^{16} \text{ cm}^{-3}$ at a depth of 6.0 microns. The drift region **202** may form an abrupt non-rectifying junction with the substrate **200**.

Conventional selective etching techniques may then be performed using a first etching mask (not shown) to define a plurality of parallel stripe-shaped trenches **204** in the drift region **202**. Trenches **204** having other shapes may also be used. For example, each pair of adjacent trenches

5 **204** may represent opposing sides of a respective ring-shaped trench. These trenches **204** may have a depth D_t of 5 microns, for example. Adjacent trenches **204** define drift region mesas **202b** therebetween, with the width W_m of each mesa **202b** controlled by the spacing between the adjacent trenches **204**. As illustrated by FIG. 9B, a thin thermal oxide layer

10 **206** may then be grown at a low temperature on the sidewalls and bottoms of the trenches **204** and on an upper surface **202a** of each of the mesas **202b**. For example, this thin oxide layer **206** may be grown for a duration of 30 minutes at a temperature of 900°C in a wet O_2 ambient. This thermal growth step may result in an oxide layer **206** having a thickness of about

15 700Å. This thin oxide layer **206** can be used to improve the interface between the sidewalls of the trenches **204** and subsequently formed regions within the trenches **204**, by removing etching related defects. The thermal budget associated with this thermal oxide growth step should be insufficient to significantly alter the graded doping profile in the drift region

20 **202**, however, the doping concentration at the surface **202a** of each mesa **202b** may increase as a result of dopant segregation. A thick conformal oxide layer **208** may then be deposited at a low temperature to produce an electrically insulating spacer on the sidewalls and bottoms of the trenches **204**. For a 65 Volt product rating, the total oxide thickness (thermal oxide

25 plus deposited oxide) may be 3000Å.

Referring now to FIG. 9C, a conformal polysilicon layer **210** may then be deposited using a low temperature CVD process. The thickness of this layer should be sufficient to fill the trenches **204**. The polysilicon layer **210** may be in-situ doped (e.g., with phosphorus) so that a low sheet resistance

30 of 10 ohms/square is achieved. As illustrated by FIG. 9D, the deposited polysilicon layer **210** may then be etched back using conventional etching

techniques. The duration of this etching step may be sufficiently long that the polysilicon regions **210a** within each trench **204** are planar with the upper surfaces **202a** of the mesas **202b**. This etch back step may be performed without an etching mask. Referring now to FIG. 9E, another etching step may then be performed with a second mask (not shown) in order to selectively remove the oxide over the mesas **202b**, but preserve the oxide within field oxide regions (not shown) that may be located around a periphery of the drift region **202**. This second mask may comprise a photoresist layer that has been patterned to define an etching window that is within a border of an outside trench (not shown) that surrounds an integrated power device containing a plurality of the illustrated power devices as unit cells.

As illustrated by FIG. 9F, a thin pad oxide layer **212** is then grown as a screening oxide over the exposed upper surfaces of the mesas **202b**. This thin pad oxide layer **212** may have a thickness of about 250Å. This thin pad oxide layer **212** may be grown for a duration of 10 minutes at a temperature of 900°C in a wet O₂ ambient. Transition region dopants **214** of first conductivity type may then be implanted using a blanket implant step. In particular, transition regions having vertically retrograded doping profiles therein relative to the upper surface **202a** may be formed by implanting phosphorus dopants at an energy level of 200 keV and at a preferred dose level of $5 \times 10^{12} \text{ cm}^{-2}$. This energy level of 200 keV and dose level of $5 \times 10^{12} \text{ cm}^{-2}$ may result in an N-type transition region having a peak implant depth (N_{PID}) of about 0.25-0.3 microns and a peak dopant concentration of about $1.3 \times 10^{17} \text{ cm}^{-3}$.

Referring now to FIG. 9G, the pad oxide layer **212** is then removed and in its place a gate oxide layer **216** having thickness of about 500Å may be formed. This gate oxide layer **216** may be provided by performing a thermal oxidation step in a wet O₂ ambient for a duration of 20 minutes and at a temperature of 900°C. A blanket polysilicon layer **218** is then deposited and patterned using a photoresist mask layer **220** (third mask),

to define a plurality of gate electrodes **218**. A sequence of self-aligned implant steps are then performed. In particular, highly doped self-aligned shielding regions of second conductivity may be formed in the transition region by implanting shielding region dopants **222** (e.g., boron) at an energy level of 100 keV and at a dose level of $1 \times 10^{14} \text{ cm}^{-2}$. After thermal treatment, these energy and dose levels may ultimately result in a shielding region having a peak boron concentration of about $5 \times 10^{18} \text{ cm}^{-3}$ at a depth of about 0.3 microns, assuming a characteristic diffusion length of about 0.1 microns. These shielding region dopants **222** are preferably implanted using both the gate electrodes **218** and the mask layer **220** as an implant mask. Self-aligned base regions of second conductivity type may also be formed in the shielding regions by implanting base region dopants **224** (e.g., boron) at an energy level of 50 keV and at a dose level of $3 \times 10^{13} \text{ cm}^{-2}$. The locations of peak concentrations of the shielding region dopants **222** and base region dopants **224** within the mesas **202b**, are represented by the reference characters "+". The peak concentration of the shielding region dopants may equal $3 \times 10^{18} \text{ cm}^{-3}$, at a depth of 0.25-0.3 microns. This depth preferably matches the depth of the peak of the transition region dopants.

Referring now to FIG. 9H, the mask layer **220** may be removed and then a drive-in step may be performed at a temperature of about 1000°C and for a duration of about 60 minutes to define self-aligned base regions **226** (shown as P), self-aligned shielding regions **228** (shown as P+) and the transition regions **230** (shown as N). This drive-in step, which causes lateral and downward diffusion of the implanted base, shielding and transition region dopants, may provide the highest thermal cycle in the herein described method. If the uniform and graded doping profile in the drift region is significantly altered during this step, then the initial drift region doping profile may be adjusted to account for the thermal cycle associated with the drive-in step. As illustrated by FIG. 9H, the implant energies and duration and temperature of the drive-in step may be chosen so that the

depth of the P-N junction between the P+ shielding region **228** and the drift region **202** is about equal to the depth of the non-rectifying junction between the transition region **230** and the drift region **202**, however, unequal depths may also be used. The depth of the P-N junction may
5 equal 0.7 microns.

Referring now to FIG. 9I, source region dopants **232** of first conductivity type are then implanted into the base regions **226**, using the gate electrodes **218** as an implant mask. The source region dopants **232** may be implanted at an energy level of 40 keV and at a dose level of
10 $2 \times 10^{14} \text{ cm}^{-2}$. As illustrated by FIG. 9J, the implanted source region dopants (shown by reference character "-") may then be driven-in at a temperature of 900°C and for a duration of 10 minutes, to define N+ source regions **233**. This implant step may be performed using the gate electrodes **218** and
15 fourth photoresist mask (not shown) as an implant mask. The fourth photoresist mask may be patterned to define the locations of shorts to the P-base region in a third dimension relative to the illustrated cross-section (not shown). Conventional insulator deposition, sidewall spacer formation and patterning steps may then be performed to define a plurality of insulated gate electrodes **234**. These steps may also be performed to
20 define contact windows to the source regions, the P-base regions, the polysilicon in the trenches and the gate electrodes. The insulating regions **206/208** lining upper sidewalls of the trenches may also be selectively etched back to expose sidewalls of the source, base and shielding regions. The presence of this etch back step may eliminate the need to define
25 shorts to the P-base region, using the fourth photoresist mask, and therefore may result in an increase in the forward on-state conduction area for a given lateral unit cell dimension. As illustrated by FIG. 9K, conventional front side metallization deposition and patterning steps may also be performed to define a source contact **238** and gate contact (not
30 shown). As illustrated, the source contact **238** extends along the upper sidewalls of the trenches **204** and contacts the exposed portions of the

source, base and shielding regions. The backside of the substrate **200** may also be thinned and then conventional backside metallization steps may be performed to define a drain contact **236**.

5 Vertical power devices according to still further embodiments of the present invention represent improvements over conventional double-diffused power MOSFETs. As illustrated by FIG. 12, a conventional unit cell of a double-diffused MOSFET **300** includes a semiconductor substrate **302** having a drift region **304** therein. An upper portion **304a** of the drift region **304** that extends adjacent an upper surface of the substrate **302** may be more highly doped and may have a downward sloping vertical
10 doping profile as illustrated on the right side of FIG. 12. As will be understood by those skilled in the art, the upper portion **304a** of the drift region **304** may be referred to as a JFET neck region. The illustrated doping profile represents the vertical concentration of first conductivity type dopants in the substrate **302**, taken along line 12A-12A'. This upper
15 portion **304a** of the drift region **304** may be formed by implanting N-type dopants at a relatively low energy level into the upper surface of the substrate **302** and then annealing the substrate **302** so that the implanted dopants diffuse downward from a peak concentration at the surface. The
20 peak concentration is preferably at the surface because the lateral distance between each pair of P-type base regions **314** is narrowest at the surface of the substrate **302** and this typically results in strong JFET action. The drift region **304** may be formed by epitaxially growing an in-situ doped semiconductor layer on a highly doped drain region **306** (e.g., N+ wafer). A
25 drain electrode **308** may be formed in ohmic contact with the drain region **306** on a lower surface of the substrate **302**.

The MOSFET **300** also includes an insulated gate electrode **318** that is spaced from the upper surface of the substrate **302** by a gate oxide layer **320**. The insulated gate electrode **318** is also isolated from an adjacent
30 source electrode **310** by an insulating capping layer **322**. The pair of P-type base regions **314** may be formed as stripe-shaped regions that extend

in a third dimension (not shown) and parallel to the gate electrode **318**. Highly doped base contact regions **312** (shown as P+) are also provided. These contact regions **312** extend through the base regions **314** and into the drift region **304**, as illustrated. The base regions **314** may be formed by
5 implanting base region dopants into the substrate **302** using the gate electrode **318** as an implant mask. An annealing step is then performed to at least partially drive-in the implanted base region dopants. After this annealing step, source region dopants may be implanted into the substrate **302** using the gate electrode **318** again as an implant mask. A short
10 duration annealing step may then be performed to simultaneously drive-in the implanted base region and source region dopants. Following this second annealing step, a masking layer may be deposited on the substrate **302** and patterned to define openings therein. A high concentration of base contact region dopants are then implanted at a relatively high energy level
15 through the openings and into the center of each base region **314**. A third annealing step may then be performed to drive the base contact region dopants laterally and vertically into the substrate **302**. The use of P+ base contact regions **312** can inhibit parasitic bipolar transistor action and improve safe-operating-area and ruggedness by reducing the effective
20 base resistance underneath each source region **316**. An additional power device (e.g., MOSFETs, IGBTs) that utilizes a buried P-type layer, which is self-aligned to the gate electrode, is disclosed in U.S. Patent No. 5,879,967 to Kim entitled "Methods of Forming Power Semiconductor Devices Having Latch-Up Inhibiting Regions." Like the device of FIG. 12, the buried P-type
25 layer in the device disclosed in the '967 patent does not extend beyond the P-base region in the lateral direction.

Referring now to FIGS. 13 and 15, MOSFET devices **400** and **400'** according to additional embodiments of the present invention have laterally
30 extending base shielding regions **412** that are highly doped and operate to shield or protect respective base regions **414** by significantly suppressing P-base reach-through effects when the MOSFET **400** is blocking reverse

voltages and by causing reverse current to flow through the base shielding regions **412** instead of the base regions **414**. As described more fully hereinbelow with respect to FIGS. 14A-14G, this suppression of P-base reach-through enables a reduction in the channel length of the device **400**.

5 In particular, the MOSFET **400** of FIG. 13 includes a semiconductor substrate **402** having a drift region **404** of first conductivity type therein that extends on and forms a nonrectifying junction with an underlying drain region **406** (shown as N+). As illustrated by FIG. 14A, the drift region **404** may be formed by epitaxially growing a uniformly or nonuniformly doped
10 epitaxial layer on an underlying highly doped substrate. A transition region **424** of first conductivity type is also provided adjacent an upper surface **402a** of the substrate **402**. The transition region **424** may be formed during the step of epitaxially growing the drift region **404**. Alternatively, the transition region **424** may be formed by implanting first conductivity type
15 dopants at a relatively high energy level into the upper surface **402a** of the substrate **402** so that a retrograded doping profile, which peaks at a first depth relative to the upper surface **402a**, can be achieved therein after an appropriate annealing step is performed. This retrograded doping profile, which is similar to the profile illustrated by FIG. 8A, is illustrated on the right
20 side of FIGS. 13 and 14A. The step of implanting the transition region dopants may be preceded by the step of forming a masking layer having an opening therein that defines an active portion of the substrate **402**.
Alternatively, a field oxide isolation region (not shown) may be formed on an inactive portion of the upper surface of the substrate **402** and then the
25 transition region dopants may be implanted using the field oxide isolation region as an implant mask.

The MOSFET of FIG. 13 also includes an insulated gate electrode **418** on the upper surface **402a**. The gate electrode **418** is spaced from the upper surface **402a** by a gate oxide layer **420**. The insulated gate
30 electrode **418** is also isolated from an adjacent source electrode **410** by an insulating capping layer **422**. The gate electrode may be stripe-shaped or

patterned as a serpentine, annular (e.g., ring, hex) or other similar shape.

Base regions **414** of second conductivity type (shown as P-type) are provided in the substrate **402** and these base regions **414** are preferably self-aligned to respective ends of the insulated gate electrode **418**. These
5 base regions **414** extend opposite the insulated gate electrode and support inversion-layer channels during forward on-state conduction. Source regions **416** of first conductivity type are provided in the base regions **416** and may be sufficiently wide to extend laterally into the more highly doped base shielding regions **412** as well. The source regions **416** extend
10 laterally underneath respective ends of the gate electrode **418**. The lateral distance between an end of each source region **416** and opposing edge of the transition region **424** define the length of the channel of the MOSFET.

The base shielding regions **412** extend underneath the base regions **414** and have a greater lateral extent relative to the base regions **414**. As
15 illustrated by FIG. 13, the base regions **414** form respective P-N junctions with opposing sides of an upper portion of the transition region **424** extending adjacent the upper surface **402a** and the base shielding regions **412**, which are more highly doped than the base regions **414**, extend laterally towards each other and constrict a neck of an upper portion of the
20 transition region **424** to a minimum width at a second depth relative to the first surface. This second depth preferably corresponds to the depth at which the dopant concentration in the base shielding regions **412** peaks. The second depth is preferably equal to about the first depth (i.e., the depth at which retrograded doping profile in the transition region **424**). Moreover,
25 the combination of the vertical doping profile in the base shielding regions **412** and the preferred retrograded doping profile in the transition region **424** facilitates complete or full depletion of the transition region **424** when the power device **400** is blocking maximum reverse voltages. Each of the illustrated base, source and base shielding regions may be separate stripe-
30 shaped regions, for example, or may be respective portions of a single respective base, source or base shielding region having an annular,

polygon or other shape. These regions may nonetheless appear as separate regions when viewed in transverse cross-section.

5 The base shielding regions **412** may be formed by implanting base shielding region dopants **412a** into the upper surface **402a** using the gate electrode **418** as an implant mask. As illustrated by FIG. 14B, the gate electrode **418** may be formed as a highly conductive layer on an underlying gate oxide insulating layer **420**. A mask **421** is then formed by depositing a masking layer on the highly conductive layer and then patterning the masking layer in the shape of the gate electrode to be formed. A selective
10 etching step may then be performed to etch through portions of the conductive layer that are not covered by the mask **421**. The gate oxide insulating layer **420** may be used as an etch stop layer. Referring now to FIG. 14C, the base shielding regions dopants **412a** are implanted at a dose level of about $2 \times 10^{14} \text{ cm}^{-2}$ and an energy level in a range between about
15 100 keV and about 150 keV into the substrate **402**, using the gate electrode **418** as an implant mask. The energy level is sufficiently high to create a peak dopant concentration at a depth in a range between about 0.3 and 0.5 microns below the upper surface **402a**. An annealing step may then be performed to drive-in the implanted base shielding region dopants
20 **412a** and thereby define intermediate shielding regions **412**. Following this annealing step, base region dopants **414a** of second conductivity type are implanted at a relatively shallow level into the upper surface **402a** of the substrate **402**, as illustrated by FIG. 14D. Another annealing step may then be performed to drive the implanted base region dopants vertically and laterally underneath the gate electrode **418** and further drive-in the
25 previously implanted and annealed base shielding region dopants **412a**. Here, the lateral extent of the base regions **414** relative to the opposing ends of the gate electrode **418** is less than the lateral extent of the base shielding regions **412**. The base shielding regions **412** operate to confine
30 the transition region **424** to a minimum width at a level corresponding to the depth at which the base shielding region dopants **412a** are implanted.

Referring now to FIG. 14E, a source implant mask (not shown) is then formed on the substrate **402**. The source implant mask may have an opening therein that exposes an upper surface of the gate electrode **418** and adjacent portions of the base regions **414**. Source region dopants **416a** are then implanted into the substrate **402** and driven-in with a short duration annealing step. As illustrated by FIG. 14F, an electrically insulating layer may be deposited on the gate electrode **418** and patterned to define an insulating capping layer **422**. Conventional metallization steps may then be performed to define a source electrode **410** on the upper surface **402a** and a drain electrode **408** on a bottom surface of the substrate **402**, as illustrated by FIG. 14G.

The MOSFET device **400** of FIG. 13 may be configured to support a flyback diode. As illustrated by FIG. 15, a Schottky rectifying junction may be provided between the source electrode **410** (acting as an anode) and another portion of the transition region **414**. This rectifying junction defines a flyback diode (shown as an electrical schematic), which may be located between adjacent unit cells within an integrated MOSFET device **400'**. In particular, the source electrode **410** may be treated as forming a Schottky rectifying junction with a "second" transition region and the base shielding regions **412** may be defined to constrict a neck of the second transition region **424**, as described above with respect to the MOSFET device **400** of FIG. 13. Moreover, to suppress reverse leakage currents across the Schottky rectifying junction, the second transition region **424** in the center of FIG. 15 may be constricted to a greater degree by the base shielding regions **412**, relative to the degree of constriction of the transition regions in the vertical MOSFET unit cells located on the left side and right side of FIG. 15.

Two-dimensional numerical simulations were performed on the vertical MOSFET of FIG. 13. For the unit cell, a gate width of 1.2 microns (when viewed in cross-section) and a gate oxide thickness of 40 nm (400 Å) was used. The drift region doping concentration was set at a level of

1.75 x10¹⁶ cm⁻³ and had a thickness of 2 microns. The depth of the base shielding regions was also set at 0.75 microns and the cell pitch was set at 3 microns. The P-base channel length was also set at 0.17 microns.

Based on these characteristics, the breakdown voltage was simulated as
5 40 Volts and a low specific on-state resistance (R_{sp}) of 0.17 milli-Ohm cm² was achieved. The specific gate charge Q_i (to V_g=4.5 Volts) was found to be 2.57x10⁻⁷ C/cm² and the specific Miller gate charge was found to be 1.1x10⁻⁷ C/cm². The Figure of Merit (FOM) corresponding to these results was 23x10⁹ (i.e., (R_{sp}xQ_i)⁻¹ = 23x10⁹). In contrast, the vertical MOSFET of
10 FIG. 12 was simulated as having a gate width of 2 microns and a gate oxide thickness of 40 nm (400 Å). The drift region doping concentration was set at a level of 1.5 x10¹⁶ cm⁻³ and had a thickness of 2 microns. The depth of the contact regions was also set at 1 microns and the cell pitch was set at 4 microns. The P-base channel length was also set at 0.5
15 microns. Based on these characteristics, the breakdown voltage was simulated as 40 Volts and a low specific on-state resistance (R_{sp}) of 0.30 milli-Ohm cm² was achieved. The specific gate charge Q_i (to V_g=4.5 Volts) was found to be 2.8x10⁻⁷ C/cm² and the specific Miller gate charge was found to be 1.5x10⁻⁷ C/cm². The Figure of Merit (FOM) corresponding to
20 these results was 12x10⁹ (i.e., (R_{sp}xQ_i)⁻¹ = 12x10⁹).

In the drawings and specification, there have been disclosed typical preferred embodiments of the invention and, although specific terms are employed, they are used in a generic and descriptive sense only and not for purposes of limitation, the scope of the invention being set forth in the
25 following claims.